

# Chapter 17

## Filters and Attenuators

1. A  $t$ -section low pass filter has series inductance 80 mH and shunt capacitance 0.022  $\mu$ F. Determine the cut-off frequency and nominal design impedance. Obtain the equivalent  $\pi$ -section.

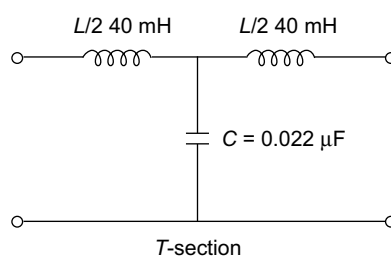
*Solution:* The cut-off frequency  $f_c$  is given by

$$\begin{aligned} f_c &= \frac{1}{\pi\sqrt{2c}} \\ &= \frac{1}{\pi\sqrt{(80 \times 10^{-3})(0.022 \times 10^{-6})}} \text{ Hz} \\ &= 7.587 \text{ kHz} \end{aligned}$$

The design impedance ( $R_0$ ) is given by

$$R_0 = \sqrt{\frac{L}{C}} = \sqrt{\frac{80 \times 10^{-3}}{0.022 \times 10^{-6}}} = 1.907 \text{ k}\Omega$$

The low-pass filters for  $T$ -section and  $\pi$ -section are shown below.



**Fig. 17.1**

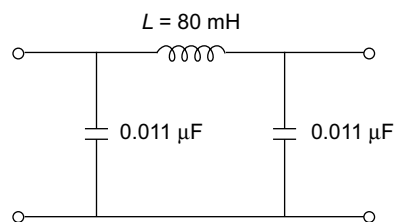


Fig. 17.2

2. Design a high-pass filter to have a design impedance of  $500\ \Omega$  and a cut-off frequency of 1 KHz.

*Solution:* Design impedance =  $500\ \Omega$

Cut off frequency = 1000 Hz

$$L = \frac{K}{4\pi f_c} = \frac{500}{4\pi \times 1000} = \frac{7}{176}\text{ H}$$

$$C = \frac{1}{4\pi K f_c} = \frac{1}{4\pi \times 500 \times 1000} = \frac{7}{44}\ \mu\text{F}$$

The  $T$  and  $\pi$  sections are shown below.

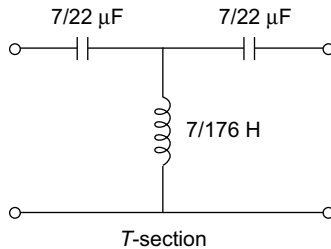


Fig. 17.3

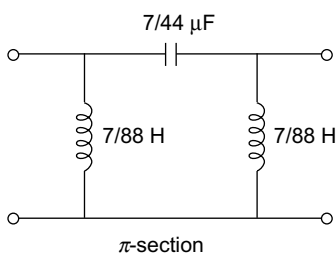


Fig. 17.4

3. Design an  $m$ -derived low-pass filter, both  $T$  and  $\pi$  sections having design impedance  $R_0 = 500\ \Omega$ , cut-off frequency  $f_c = 1500\text{ Hz}$  and infinite attenuation frequency,  $f_\infty = 2000\text{ Hz}$ .

*Solution:* From the theory of  $m$ -derived filter,

$$m = \sqrt{1 - \left(\frac{f_c}{f_\infty}\right)^2}$$

$$= \sqrt{1 - \left(\frac{1500}{2000}\right)^2} = 0.661$$

For the prototype low-pass filter, for  $f_c = 1500$  Hz and  $R_0 = 500 \Omega$ , series arm inductance is given by

$$L = \frac{R_0}{\pi f_c} = \frac{500}{\pi \times 1500} \text{ H} = 106.103 \text{ mH}$$

and shunt-arm capacitance is given by

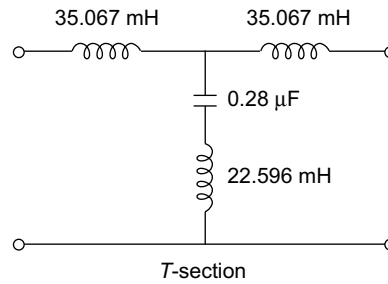
$$C = \frac{1}{\pi R_0 f_c} = \frac{1}{\pi \times 500 \times 1500} \text{ F} = 0.424 \mu\text{F}$$

*T-section:* The elements are

$$\frac{mL}{2} = \frac{0.661 \times 106.103}{2} = 35.067 \text{ mH}$$

$$mC = 0.661 \times 0.424 = 0.280 \mu\text{F}$$

$$\frac{1-m^2}{4m} L = \frac{1-(0.661)^2}{4 \times 0.661} \times 106.103 = 22.596 \text{ mH}$$



**Fig. 17.5**

In the  $\pi$ -section of  $m$ -derived low-pass filter, the values of the elements are

$$\frac{mC}{2} = \frac{0.661 \times 0.424}{2} = 0.14 \mu\text{F}$$

$$mL = 0.661 \times 106.103 = 70.134 \text{ mH}$$

$$\frac{1-m^2}{4m} C = \frac{1-(0.661)^2}{4 \times 0.661} \times 0.424 = 0.090 \mu\text{F}$$

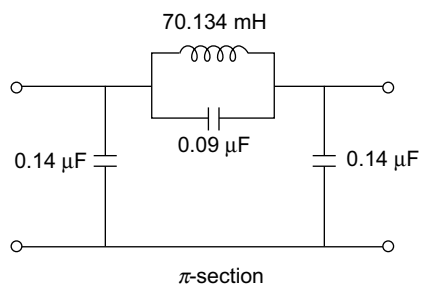


Fig. 17.6

4. Design an  $m$ -derived high-pass filter having a design impedance of  $500 \Omega$  and a cut-off frequency of 1 KHz. Take  $m = 0.2$ .

*Solution:* The constant  $K$ -filter has the values (Problem 2)

$$L = \frac{7}{176} \text{ H and } C = \frac{7}{44} \mu\text{F}$$

$$\frac{C}{m} = \frac{7}{44 \times 0.2} = \frac{35}{44} \mu\text{F}$$

$$\frac{L}{m} = \frac{7}{176 \times 0.2} = \frac{35}{176} \text{ H}$$

$$\left( \frac{4m}{1-m^2} \right) C = \frac{0.8}{0.96} \times \frac{7}{44} \times 10^{-6} = \frac{35}{264} \mu\text{F}$$

$$\left( \frac{4m}{1-m^2} \right) L = \frac{0.8}{0.96} \times \frac{7}{176} = \frac{35}{1056} \text{ H}$$

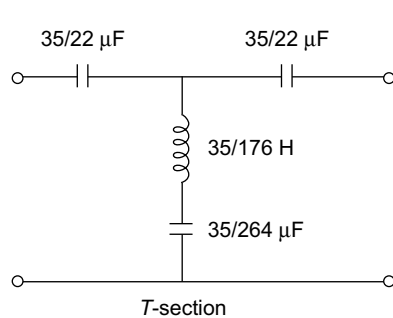


Fig. 17.7

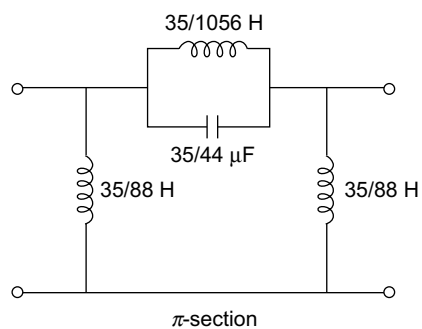


Fig. 17.8

5. Design a prototype band pass filter, both  $T$  and  $\pi$  sections having cut-off frequencies of 3000 Hz and 6000 Hz and nominal characteristics impedance of  $600 \Omega$ . Also find the resonant frequency of shunt arm of series arm.

*Solution:* The series arm impedance of a band pass filter is given by

$$L_1 = \frac{R_0}{\pi(f_2 - f_1)} = \frac{600}{\pi(6000 - 3000)} = 63.622 \text{ H}$$

$$\frac{L_1}{2} = 31.831 \text{ mH}$$

The series arm capacitance  $C_1$  is given by

$$C_1 = \frac{f_1 - f_2}{4\pi R_0 f_1 f_2} = \frac{6000 - 3000}{4\pi \times 600 \times 3000 \times 6000} = 0.022 \mu\text{F}$$

$$2C_1 = 0.044 \mu\text{F}$$

The shunt-arm inductance  $L_2$  and capacitance  $C_2$  are given by

$$L_2 = \frac{R_0(f_1 - f_2)}{4\pi f_1 f_2} = \frac{600(6000 - 3000)}{4 \times \pi \times 6000 \times 3000}$$

$$L_2 = 7.958 \text{ mH}$$

$$2L_2 = 15.916 \text{ mH}$$

$$C_2 = \frac{1}{\pi R_0 (f_1 - f_2)} = \frac{1}{\pi \times 600 \times 3000}$$

$$C_2 = 0.177 \mu\text{F}$$

Hence  $\frac{C_2}{2} = 0.0885 \mu\text{F}$

Resonant frequency is  $f_0 = \sqrt{f_1 f_2} = \sqrt{3 \times 6 \times 10^6} = 4242.7 \text{ Hz}$

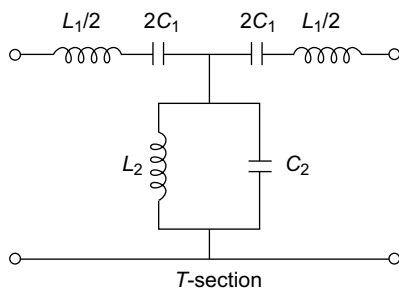


Fig. 17.9

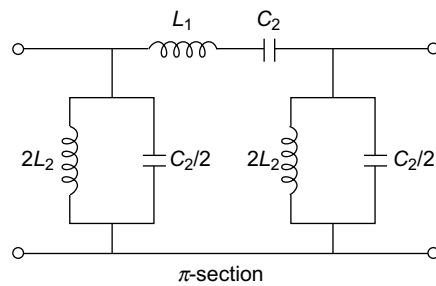


Fig. 17.10

6. Design prototype band stop filter section having cut-off frequency of 2000 Hz and 5000 Hz and a design resistance of 600  $\Omega$ .

*Solution:* The series arm inductance  $L_1$  and capacitance  $C_1$  are given by

$$L_1 = \frac{R_0(f_2 - f_1)}{\pi f_1 f_2} = \frac{600(5000 - 2000)}{\pi \times 2000 \times 5000} = 57.296 \text{ mH}$$

$$\frac{L_1}{2} = 28.648 \text{ mH}$$

$$C_1 = \frac{1}{4\pi R_0(f_2 - f_1)}$$

$$= \frac{1}{4\pi \times 600(5000 - 2000)} = 0.044 \text{ } \mu\text{F}$$

$$2C_1 = 2 \times 0.044 = 0.088 \text{ } \mu\text{F}$$

The shunt inductance  $L_2$  and capacitance  $C_2$  are given by

$$L_2 = \frac{R_0}{4\pi(f_2 - f_1)} = \frac{600}{4\pi \times (5000 - 2000)}$$

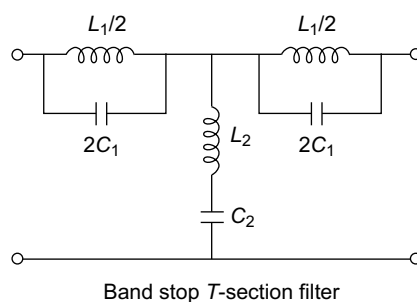
$$L_2 = 15.915 \text{ mH}$$

and

$$C_2 = \frac{f_2 - f_1}{4\pi R_0(f_1 f_2)} = \frac{5000 - 2000}{\pi \times 600 \times 2000 \times 5000}$$

$$C_2 = 0.159 \text{ } \mu\text{F}$$

The band stop  $T$ -section filter is



**Fig. 17.11**

7. An attenuator is composed of symmetrical  $\pi$ -section having series arm of  $275 \text{ } \Omega$  and shunt arm each of  $450 \text{ } \Omega$ . Find

- (i) the characteristic impedance of the network, and
- (ii) attenuation per section

*Solution:* On the symmetrical  $\pi$  alternator

$$R_1 = R_0 \left( \frac{N^2 - 1}{2N} \right)$$

$$R_1 = 275 \text{ } \Omega \text{ and } R_2 = 450 \text{ } \Omega$$

$$\therefore R_0 \left( \frac{N^2 - 1}{2N} \right) = 275 \quad (1)$$

$$R_2 = R_0 \left( \frac{N + 1}{N - 1} \right)$$

$$R_0 \left( \frac{N + 1}{N - 1} \right) = 450 \quad (2)$$

Dividing Equation (1) by (2), we get  $N$

$$N = 2.874 \text{ or } 0.348$$

If  $N = 0.348$ , it provides gain than attenuation

Thus  $N = 2.874$

$$(i) \quad R_1 = R_0 \left( \frac{N^2 - 1}{2N} \right)$$

$$R_0 = R_1 \times \left( \frac{2N}{N^2 - 1} \right) = \frac{275 \times 2 \times 2.874}{(2.874)^2 - 1}$$

$$R_0 = 217.731 \, \Omega$$

(ii) attenuation per section

$$D = 20 \log_{10} (2.874) = 9.17 \text{ dB.}$$

8. Design a symmetrical lattice attenuator to have attenuation of 20 dB and characteristic impedance of 500  $\Omega$ . What will be the equivalent  $T$ -configuration?

*Solution:*  $R_0 = 500 \, \Omega$  and  $D = 20 \text{ dB}$

$$N = \text{antilog}_{10} \left( \frac{20}{20} \right) = 10$$

In symmetrical lattice attenuator

$$R_1 = R_0 \left( \frac{N - 1}{N + 1} \right) = 500 \left( \frac{10 - 1}{10 + 1} \right) = 409.091 \, \Omega$$

$$R_2 = R_0 \left( \frac{N + 1}{N - 1} \right) = 500 \left( \frac{10 + 1}{10 - 1} \right) = 611.111 \, \Omega$$

The equivalent design elements of  $T$ -sections are

$$R_1 = R_0 \left( \frac{N - 1}{N + 1} \right) = 500 \left( \frac{10 - 1}{10 + 1} \right) = 409.091 \, \Omega$$

$$R_2 = R_0 \left( \frac{2N}{N^2 - 1} \right) = 500 \left( \frac{2 \times 10}{10^2 - 1} \right) = 101.010 \, \Omega$$

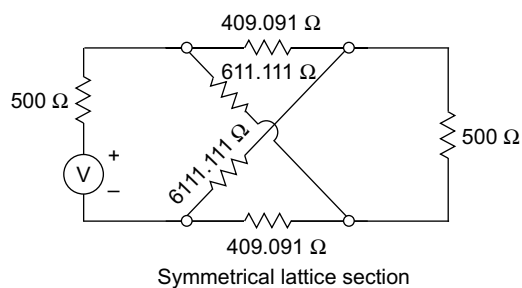


Fig. 17.12

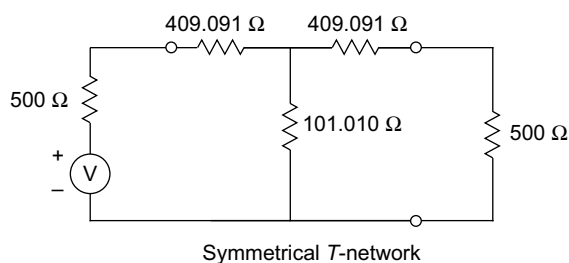


Fig. 17.13

9. The series arm of a lattice equalizer used in conjunction with a telephone line of characteristic impedance  $Z_0 = 600 \Omega$  consists of a resistor of  $400 \Omega$  in series with an inductor of  $40 \text{ mH}$ . Find the shunt of the lattice.

*Solution:* The lattice equalizer used with telephone line will be a constant resistance equalizer. The typical constant resistance lattice attenuation equalizer is shown below.

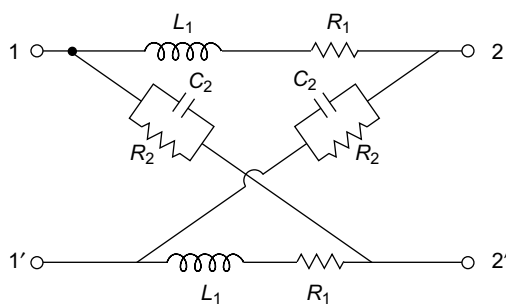


Fig. 17.14

From the given data

$$R_0 = 600 \Omega, R_1 = 400 \Omega, L_1 = 40 \text{ mH}$$

$$R_1 R_2 = \frac{L}{C} = R_0^2$$



Thus 
$$R_2 = \frac{R_0^2}{R_1} = \frac{(600)^2}{400} = 900 \, \Omega$$

$$C_2 = \frac{L}{R_0^2} = \frac{0.04}{(600)^2} = 0.111 \, \mu\text{F}$$

10. In a constant resistance lattice attenuation equalizer, having characteristic impedance of  $1 \, \Omega$ , the diagonal arm consists of a resistance  $L_2 = 1 \, \Omega$  in series with a capacitor  $C_2 = 0.05 \, \text{F}$ . Find the series arm.

*Solution:* The constant resistance lattice attenuation equalizer is shown in the figure below.

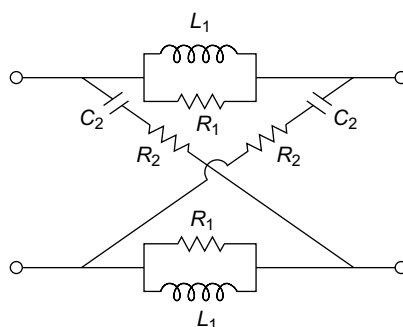


Fig. 17.15

From the given data  $R_2 = 1 \, \Omega$ ,  $C_2 = 0.05 \, \text{F}$  and  $R_0 = 1 \, \Omega$   
For a constant resistance equalizer

$$R_1 R_2 = \frac{L_1}{C_2} = R_0^2$$

Thus 
$$R_1 = \frac{R_0^2}{R_2} = \frac{(1)^2}{1} = 1 \, \Omega \text{ and } L_1 = C_2 R_0^2$$
  

$$= (0.05) (1)^2 = 0.05 \, \text{H}$$

### Objective-Type Questions

- In a broadcasting receiver, a filter circuit is used in  $IF$  circuit to limit the receiver bandwidth. The filter is a
  - band-stop filter
  - low-pass filter
  - band-pass filter
  - high-pass filter
- In the  $m$ -derived low-pass filter, the resonant frequency is to be chosen so that it is
  - above the cut-off frequency
  - below the cut-off frequency

- (c) same as that of the cut-off frequency
- (d) none of the above
- 3. A  $m$ -derived low-pass filter has  $f_\infty$  position
  - (a) above  $f_c$
  - (b) below  $f_c$
  - (c) at zero
  - (d) at infinity
- 4. A line works as
  - (a) attenuator
  - (b) LPF
  - (c) HPF
  - (d) neither of the above
- 5. Attenuation is expressed in
  - (a) decibels
  - (b) nepers
  - (c) both
  - (d) none
- 6. The two networks having impedances  $z_1$  and  $z_2$  are said to be inverse if
  - (a)  $Z_1 Z_2 = R_0^2$
  - (b)  $\frac{Z_1}{Z_2} = R_0^2$
  - (c)  $Z_1 R_0 = Z_2$
  - (d) none
- 7. Attenuation distortion in equalizers occurs due to
  - (a) non uniform attenuation against frequency
  - (b) uniform attenuation against frequency
  - (c) non-uniform attenuation against time
  - (d) uniform attenuation against time